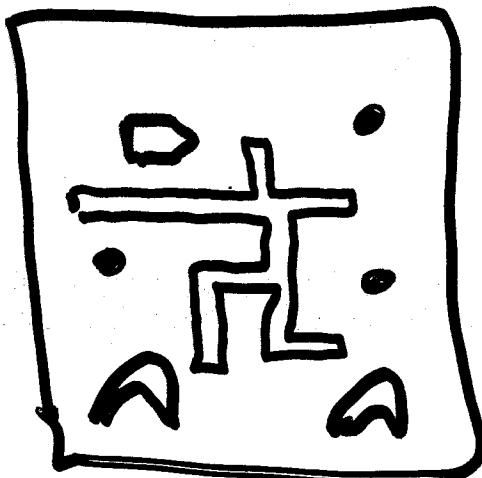


2-D Simulator



Robots
- Good
- Evil
- Prey

Food
Rewards
Obstacles

etc.

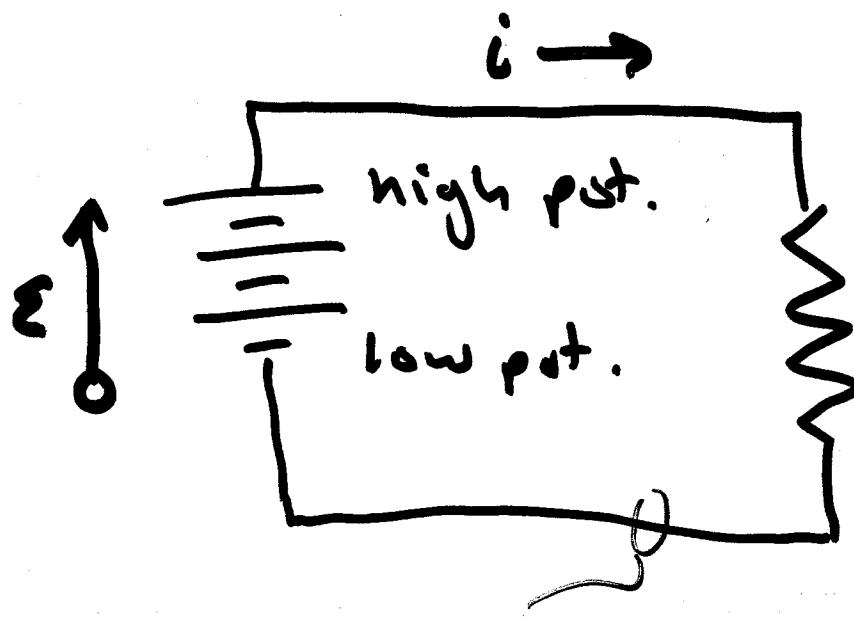
- LOTS OF FREEDOM IN DESIGN OF WORLD
- OCCUPANTS
- SENSORS
- ENVIRONMENT
- COMPLEXITY OF BEHAVIOR
- START SIMPLE
- MATLAB (SAH?)
- PUBLIC PRESENTATION
- WRITE UP

SIMULATOR PHASE II

- o Simpler logic
- o More realistic physical modeling
- o model your Tutebot.

EMF of circuits

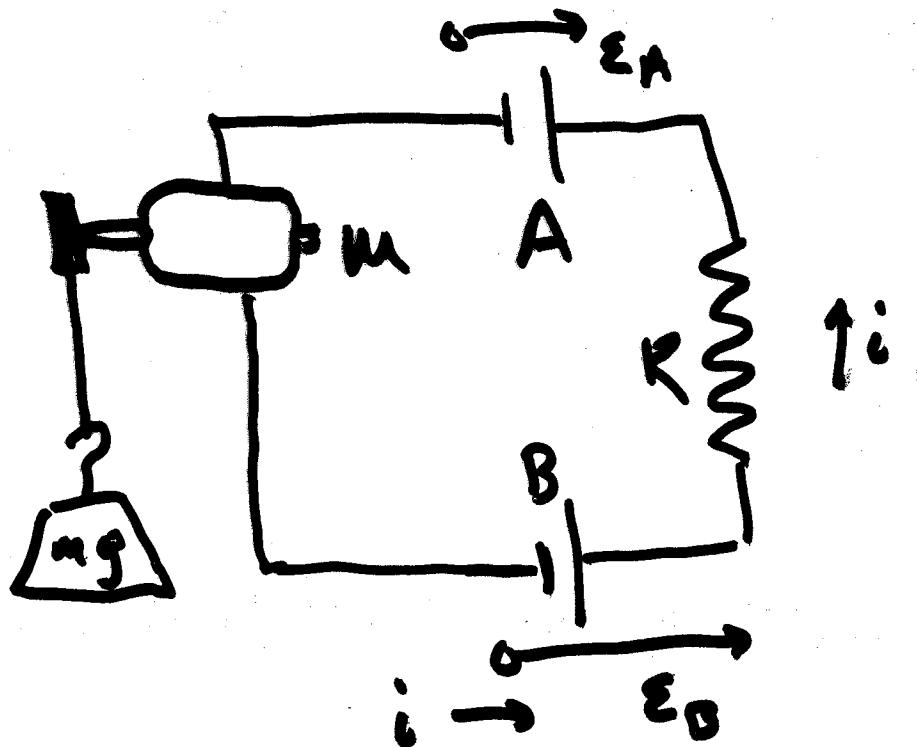
Battery or other seat of emf maintains electric potential difference between 2 points. (This p.d. moves electrons).



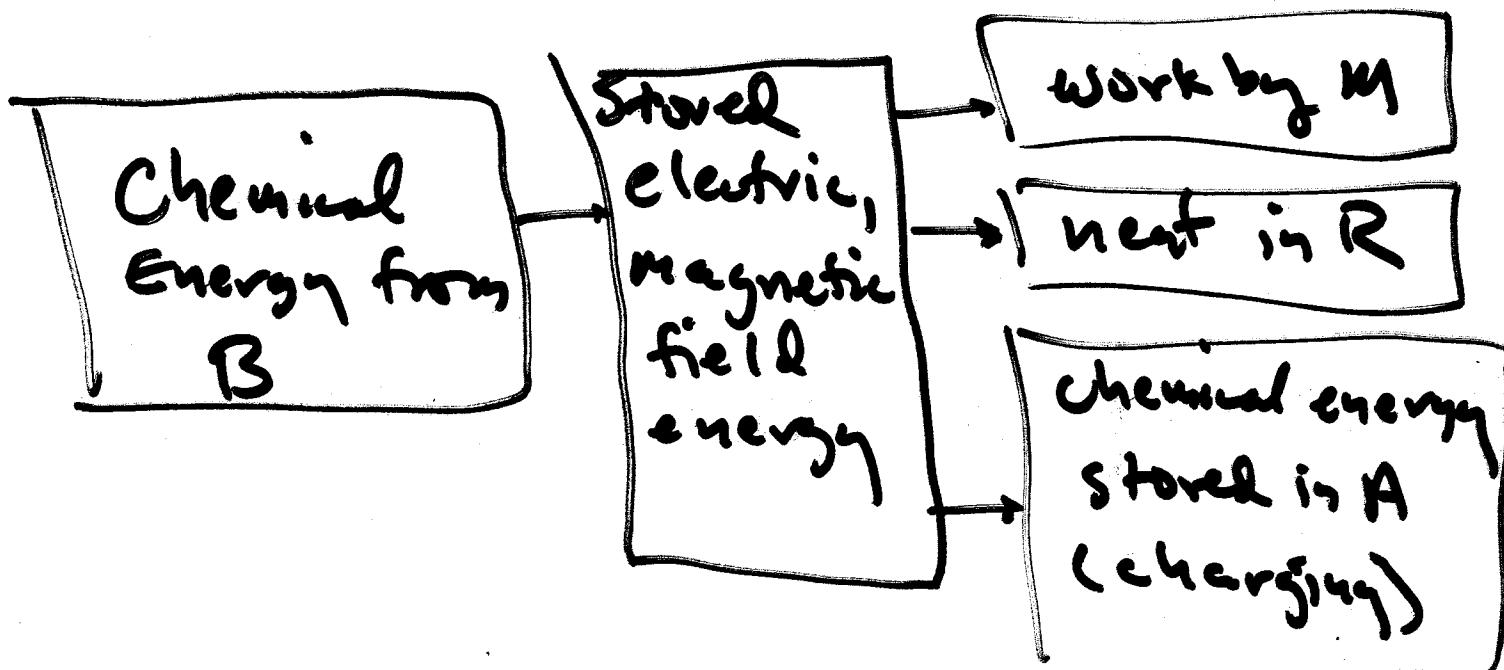
positive
charge carriers
determine
direction of
current.
(convention)

dq charge passes thru here
in time dt . seat does
 dW work to push them up
to higher potential.

$$\text{emf } \Sigma \triangleq \frac{dW}{dq} = \frac{\text{joule}}{\text{coulomb}} = \underline{\underline{\text{volt}}}$$



- Opposing batteries: B wins ($\epsilon_B > \epsilon_A$)



a "Reversible" reaction, unlike loss of energy through heating $\sqrt{2}$

joule \triangleq 1 newton-meter ; unit of work.

Newton $\triangleq \frac{\text{kg} \cdot \text{m}}{\text{sec}^2}$; unit of force

$$F = m a \quad (\text{Newton's Law})$$

$$wt = kg \frac{m}{\text{sec}^2}$$

coulombs \triangleq charge through wire in 1 second if current is 1 ampere
 $q = it$.

e $\triangleq 1.60206 \times 10^{-19}$ coul
the elementary charge.

Current Calculation

Joule heating: $i^2 R dt$ is time dt
caused by

Charge : $i dt = dq$ is time dt .

dq moves thru the seat. So...
def. of curr

$$dW = \sum dq = \sum i dt$$

this work = Joule heating by conservation
of energy, so

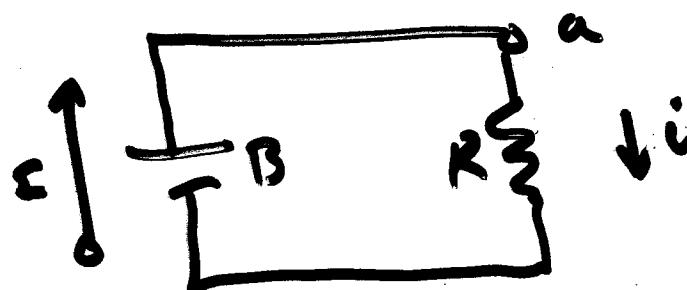
$$\sum i dt = i^2 R dt$$

$$\sum i = i^2 R$$

$$\sum = IR$$

$$i = \frac{\sum}{R}$$

this relation $i = \frac{\epsilon}{R}$ is a consequence of fact that the algebraic sum of changes in potential encountered in complete traversal of a circuit must be zero.



$$V = iR$$

thru R, $-iR$
(ohm's law)
R measured in
ohms.

Negative sign trip is at higher pot.
Traversing B from bottom to top, i is clockwise.
so if initial potential is V_a ,

$$V_a - iR + \epsilon = V_a$$

$$-iR + \epsilon = 0$$

energy argument, potential argument.

Kirchhoff's 2nd Law: "Loop theorem"

Sum of potential changes around
a circuit = 0.

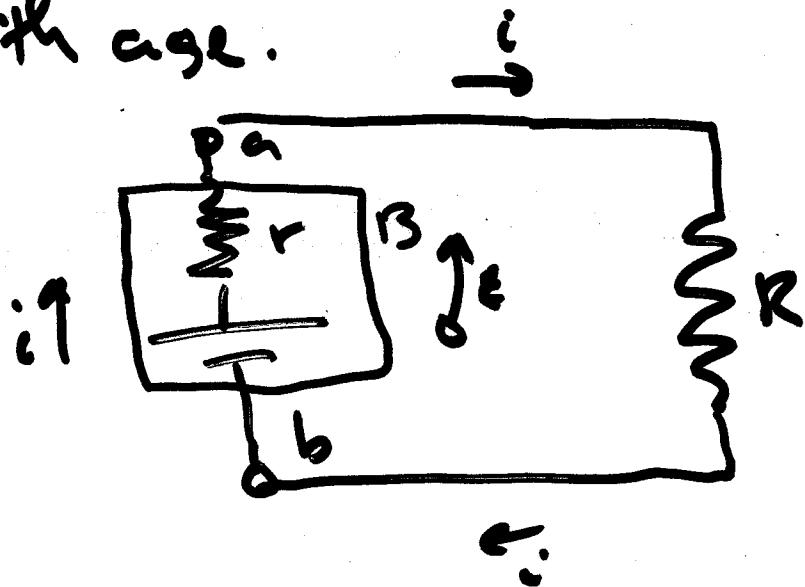
$R = \frac{V}{i}$. Resistors turn current
into voltage and vice versa.

Traverse iR in direction of current change
in potential is $-iR$. In opposite
direction is $+iR$.

Traverse ϵ in direction of cut,
change in potential is $+\epsilon$. In opposite
direction is $-\epsilon$.

E.G. #1.

Batteries (and all emf seats) have internal resistance. In batts, it grows with age.



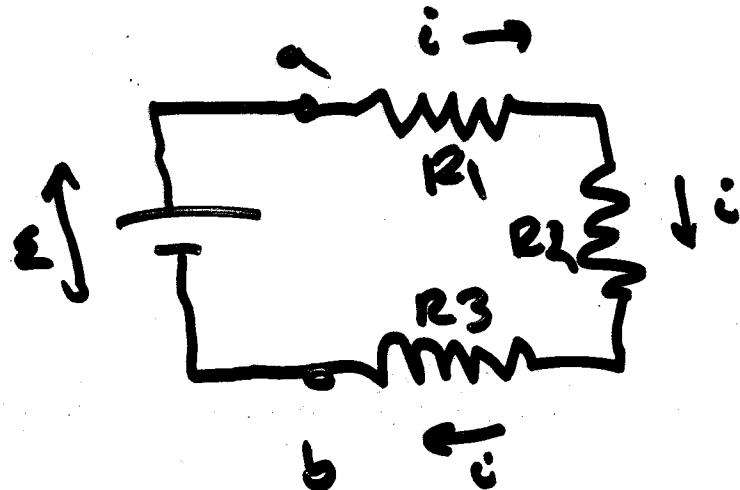
Start at b, go clockwise:

$$V_b + \epsilon - ir - iR = V_b$$

$$+ \epsilon - ir - iR = 0$$

$$i = \frac{\epsilon}{R+r}$$

Eg #2 Resistors in Series.



clockwise from a:

$$-iR_1 - iR_2 - iR_3 + E = 0$$

$$i = \frac{E}{R_1 + R_2 + R_3}$$

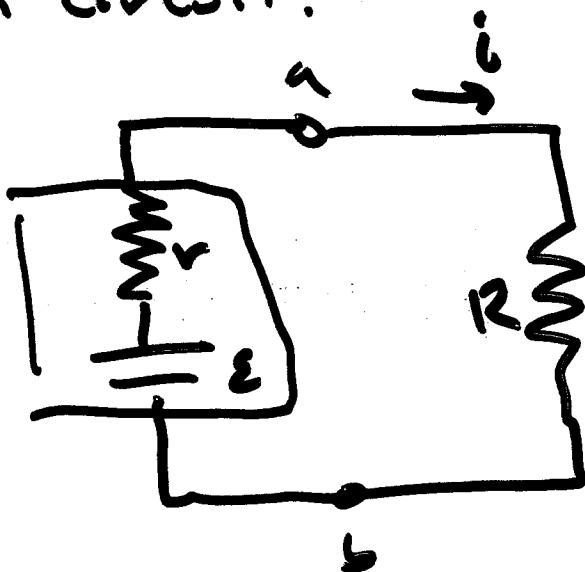
so equivalent resistance

$$R = R_1 + R_2 + R_3.$$

(They add!)

POTENTIAL DIFFERENCE

How compute the pot. diff. of 2 points in a circuit?



What is V_{ab}

$$= V_a - V_b$$

for ϵ, r, R ?

$$V_b + iR = V_a \quad V_{ab} = V_a - V_b = +iR$$

a is more positive

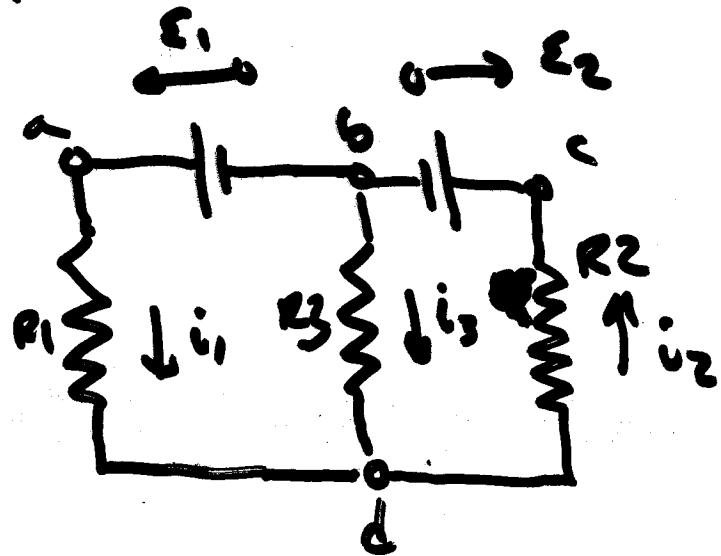
potential diff. = iR .

We computed i before : $i = \frac{\epsilon}{R+r}$

$$\text{so } V_{ab} = \epsilon \frac{R}{R+r}$$

So pot. diff. is algebraic Σ of
pot. changes from a to b.
(partial loop theorem). All paths yield same $\sqrt{19}$

MULTI-LOOP CIRCUITS



b & d are junctions connected by 3
branches. (bad, bd, bacd).

given E_s and R_s , what are i_s ?

label i, directions arbitrarily chosen.
(i3?!)

Charge does not accumulate or
drain from d. so if + is coming to
d, -is leave it,

$$i_1 + i_3 - i_2 = 0 \quad (1)$$

Junction theorem: Kirchhoff's 1st Law

\sum of currents at any junction = 0
(charge conservation)

- apply this rule at b, get some eqn.
so if we want to solve for 3 i's, need
2 more eqns.
- we have several loops with \neq
currents, so can use loop theorem.

left : $\varepsilon_1 - i_1 R_1 + i_3 R_3 = 0 \quad (2)$

right : $-i_3 R_3 - i_2 R_2 - \varepsilon_2 = 0 \quad (3)$

another : no new relationships

so solve ①, ②, ③.

e.g. $i_1 = \frac{\varepsilon_1 (R_2 + R_3) - \varepsilon_2 R_3}{R_1 R_2 + R_2 R_3 + R_1 R_3}$

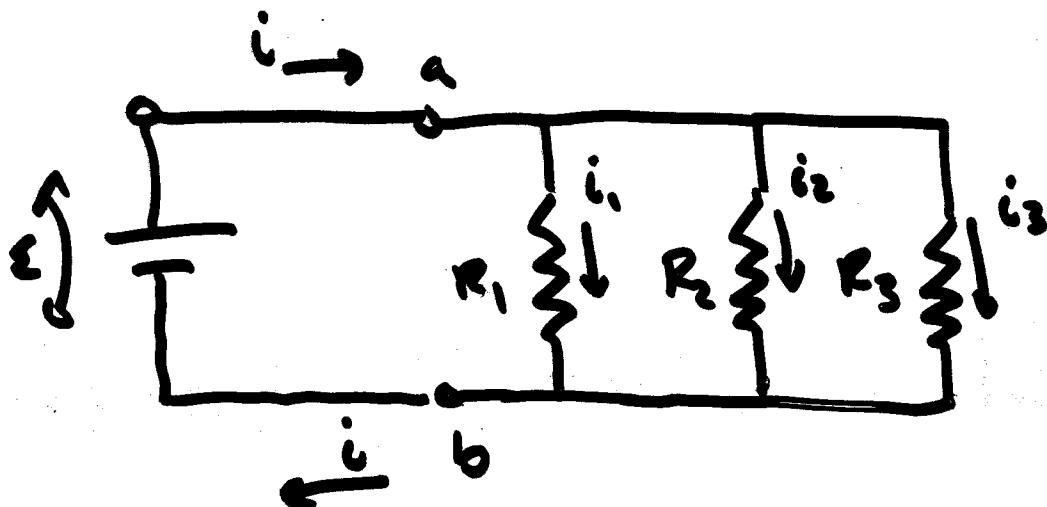
i₂ = always -1

check
currents like
 $R_3 = 0$

11

E.G.

RESISTORS IN PARALLEL



What is equivalent resistance $\overset{R}{\text{of}} R_1, R_2, R_3?$
($>?$ $<?$)

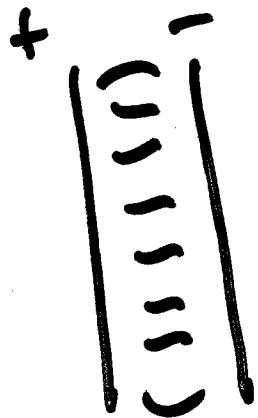
Let V be the potential diff. between
a, b. Then

$$i_1 = \frac{V}{R_1}, \quad i_2 = \frac{V}{R_2}, \quad i_3 = \frac{V}{R_3}$$

Junction theorem at a:

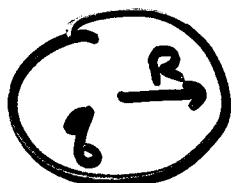
$$\textcircled{1} \quad i = i_1 + i_2 + i_3 = V \left(\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right)$$

$$\textcircled{2} \quad i = \frac{V}{R}, \quad \text{or} \quad \frac{1}{R} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$$



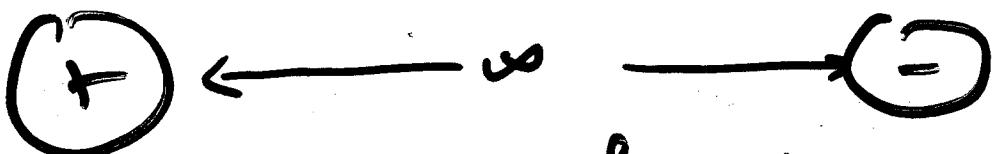
CAPACITANCE

CHARGED CONDUCTING SPHERE



has a potential:

$$V_+ = \frac{1}{4\pi\epsilon_0} \frac{q}{R}$$



Twice the potential difference



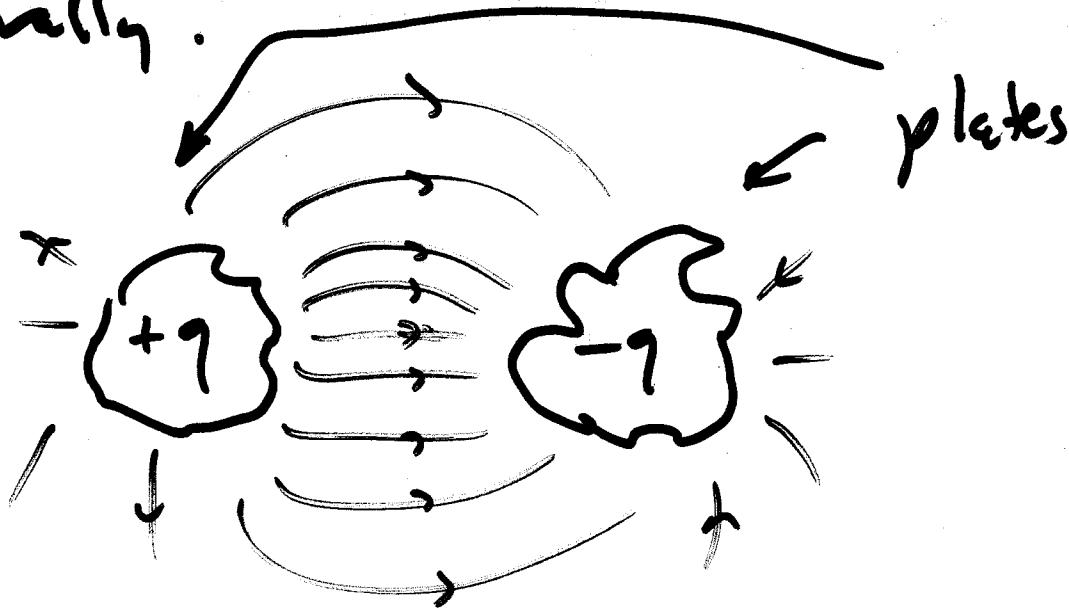
Spheres get bigger,
closer...

At 0 distance $\sqrt{13}$

$$C = \frac{q}{V}$$

$C: \text{farad} = \frac{\text{coul}}{\text{volt}}$.
 μf
 μaf in practice

generally.



A capacitor.

Practically, one type is foil:

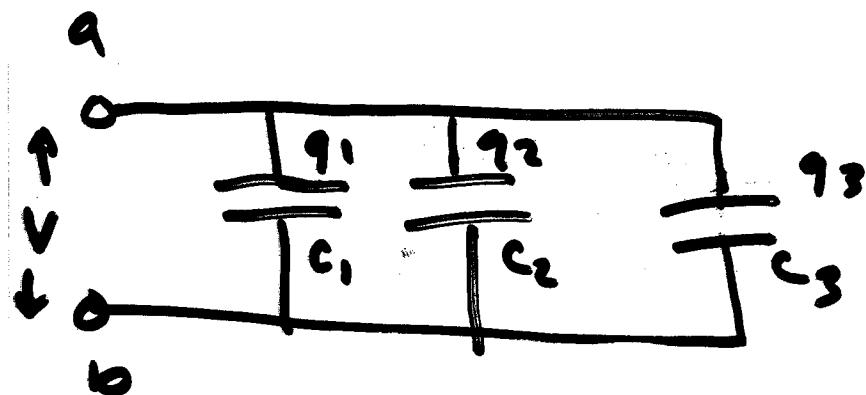


- insulator
- positive
- negative

CAPACITORS in Practice

$$C = \frac{q}{V} \quad q = CV \quad \text{or} \quad \frac{dq}{dt} = C \frac{dV}{dt} = i$$

E.G. Capacitors in Parallel



pot. difference same for all: all upper plates connected to a, lower to b.

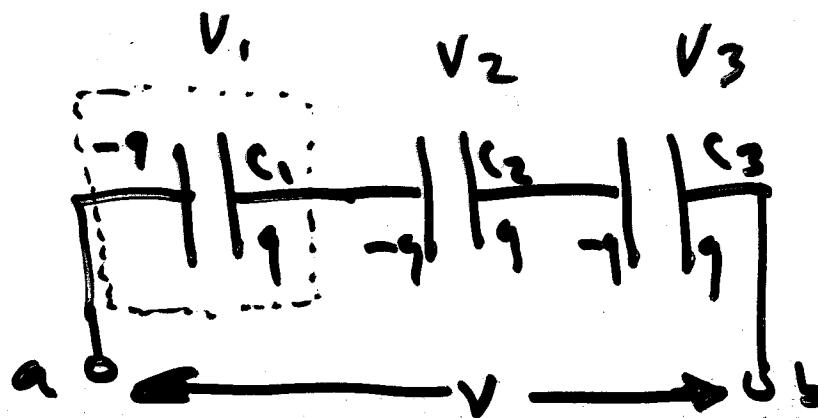
$$q_1 = C_1 V \quad q_2 = C_2 V \quad q_3 = C_3 V$$

$$q = q_1 + q_2 + q_3 = (C_1 + C_2 + C_3) V$$

$$C = \frac{q}{V} = C_1 + C_2 + C_3$$

(like series resistors)

E.G. Capacitors in Series



q must be same on all plates. Initially no charge, battery just moves it around.
charge can't leave the [] area.

$$V = \frac{q}{C_1} \quad V_2 = \frac{q}{C_2} \quad V_3 = \frac{q}{C_3}$$

$$V = V_1 + V_2 + V_3 \quad (\text{series})$$

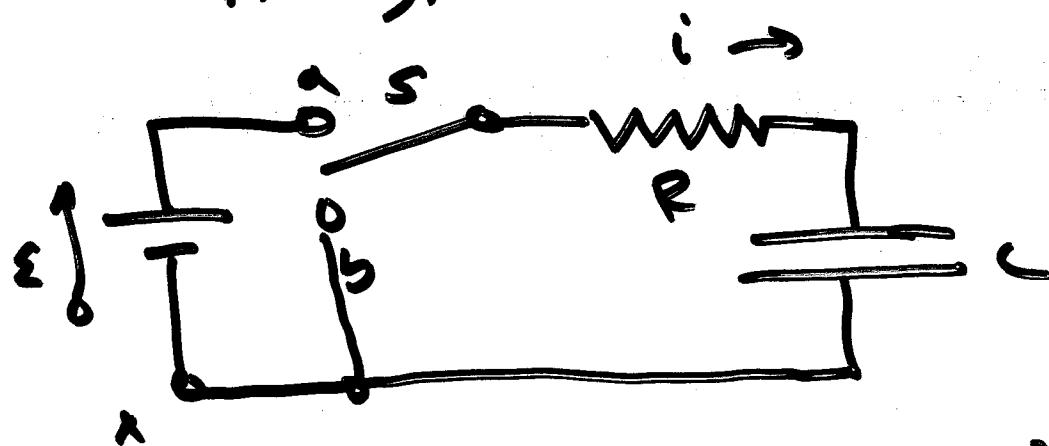
$$= q \left(\frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3} \right)$$

like $\frac{1}{R}$ resistors.

$$C = \frac{q}{V} = \frac{1}{\frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3}} \quad \text{or} \quad \frac{1}{C} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3}$$

RC circuit

Capacitors \Rightarrow time-varying currents!
 (physics is energy storage in electric field).



throw S to a: then what?

Work is Joule heat + charging C.

$$\Sigma dq = i^2 R dt + \underbrace{d\left(\frac{q^2}{2C}\right)}$$

$$\Sigma dq = i^2 R + \frac{1}{C} dq \quad \begin{matrix} \text{physics we haven't} \\ \text{seen} \end{matrix}$$

$$\Sigma \frac{dq}{dt} = i^2 R + \frac{1}{C} \frac{dq}{dt} \quad \frac{dq}{dt} = i, \text{ so}$$

$$\Sigma = iR + \frac{q}{C} \quad \boxed{\text{or from loop theorem}}$$

$$\epsilon = iR + \frac{q}{C} ; i = \frac{dq}{dt}$$

so

$$\epsilon = R \frac{dq}{dt} + \frac{q}{C} \quad \text{diff. eq.}$$

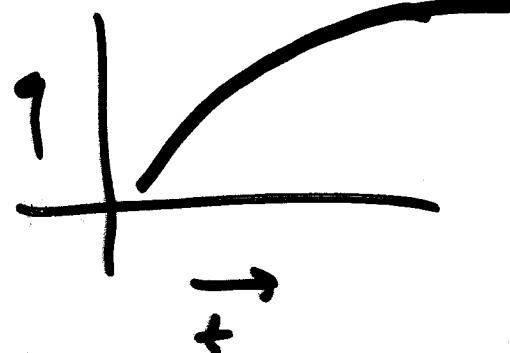
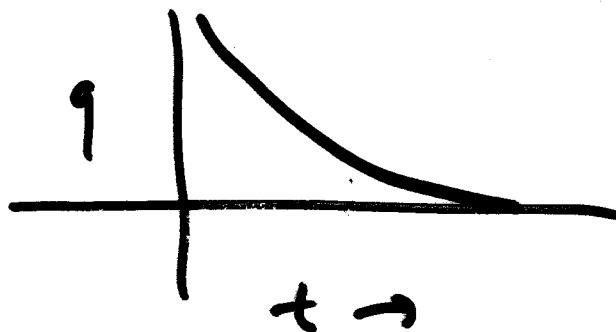
Just a slight generalization ↗

$$\begin{aligned}\dot{x} + ax &= 0 \\ \dot{x} &= -ax \\ -ae^{-at} &= -ae^{-at}\end{aligned}$$

the secret.

if $x = e^{-at}$

Solutions to 1st order linear ordinary des
are of form e^{-kt}



* CHECK

$$\Sigma = R \frac{dq}{dt} + \frac{q}{C}$$

$$\text{claim } q = C\Sigma(1 - e^{-t/RC})$$

$$\frac{dq}{dt} = i = -\frac{C\Sigma}{RC} e^{-t/RC}$$

$$\text{recall } \Sigma = R \frac{dq}{dt} + \frac{q}{C}$$

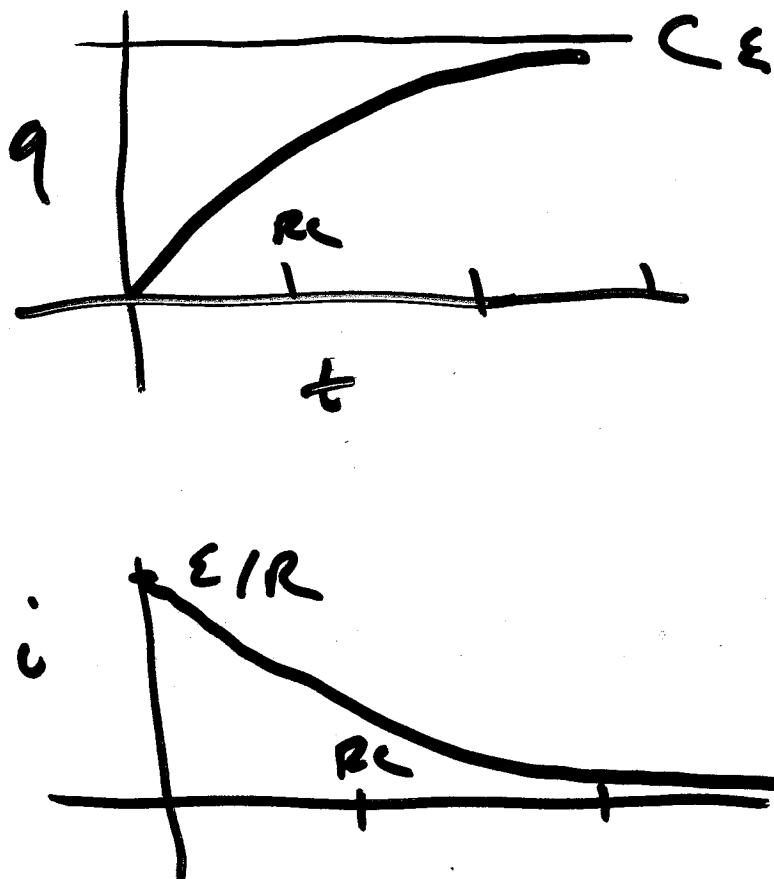
$$\Sigma = R \frac{\Sigma}{RC} e^{-t/RC} + \frac{C\Sigma(1 - e^{-t/RC})}{C}$$

$$\Sigma = \Sigma e^{-t/RC} + \Sigma (1 - e^{-t/RC}) = \Sigma \blacksquare$$

So the general soln (can be guessed) is

$$q = C\varepsilon(1 - e^{-t/RC})$$

(can be checked). *



What happens
when we
throw the
switch?

RC has dimensions of time : the
capacitive time constant.

Time it takes to charge to $(1 - e^{-1}) = 63\%$. $\sqrt{1}$

Now throw switch to b. after $t \gg RC$



Law theorem:

$$iR + \frac{q}{C} = 0$$

$$i = \frac{dq}{dt}, \text{ so dif. eq. is } R \frac{dq}{dt} + \frac{q}{C} = 0$$

soln is $q = q_0 e^{-\frac{t}{RC}}$ with q_0 the initial charge $= C\varepsilon$.

Current during discharge $i = \frac{dq}{dt} =$

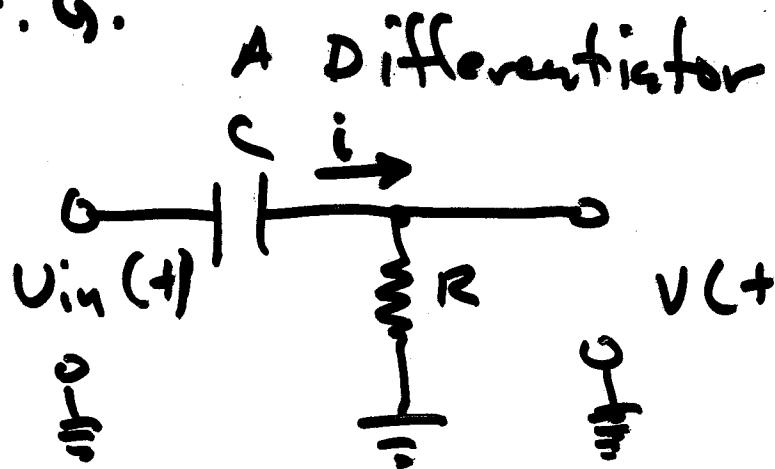
$$= -\frac{q_0}{RC} e^{-\frac{t}{RC}}$$

so current is reversed from direction in picture. Since $q_0 = C\varepsilon$

$$i = -\frac{\varepsilon}{R} e^{-\frac{t}{RC}}$$

so initially current is $-\frac{\varepsilon}{R}$ which makes sense since Cap is charged to full p.t. diff $\frac{\varepsilon_1}{\varepsilon_2}$.

E.G.



Ground.
what to measure
V's w.r.t.

Voltage across C is $V_{in} - V$.

$$I = C \frac{dV}{dt}$$
 (capacitor). definition

$$i = C \frac{dV}{dt}$$

$$I = C \frac{dV}{dt} (V_{in} - V)$$

$$I = \frac{V}{R} \text{ also}$$

if choose $\frac{dV}{dt} \ll \frac{dV_{in}}{dt}$

{ make RC small.
If R too small
you load the input

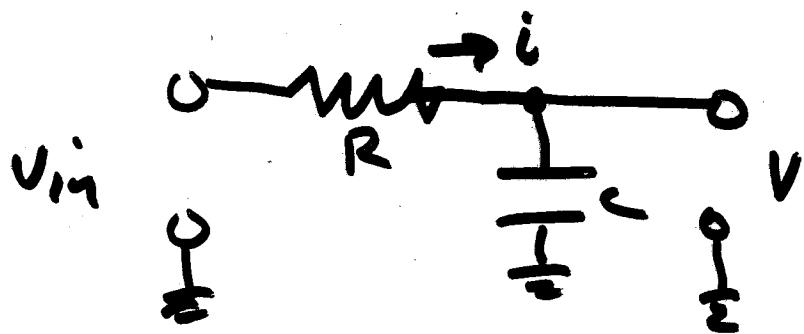
$$C \frac{dV_{in}}{dt} \approx \frac{V}{R}$$

$$V(t) = RC \frac{dV_{in}}{dt}$$

Ans!



E.G. An INTEGRATOR



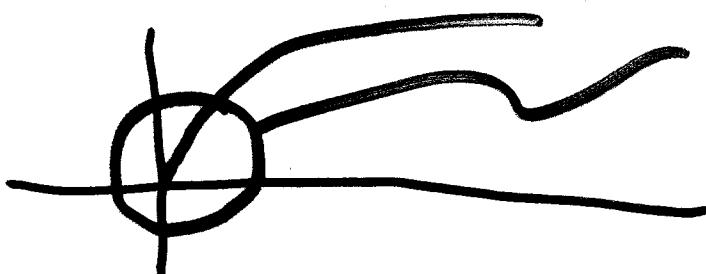
Voltage across $R = V_{in} - V$, so

$$I = \frac{V_{in} - V}{R} \quad \text{if } V \ll V_{in} (\text{with large } RC)$$

$$C \frac{dV}{dt} \approx \frac{V_{in}}{R} \approx I$$

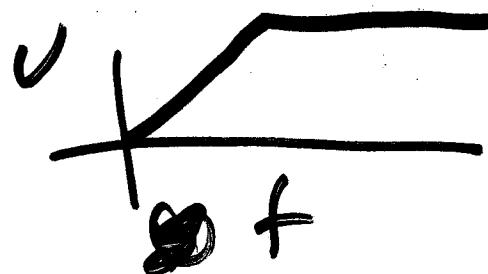
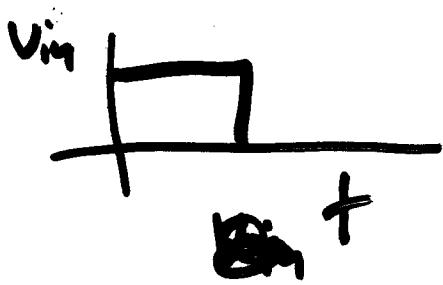
$$V(+)=\frac{1}{RC} \int_0^t V_{in}(t) dt + \text{const.}$$

Note: $V \ll V_{in} \equiv I \ll V_{in}$. we have perfect integrator if we had a current source. A large voltage across a large R approximates a current source.

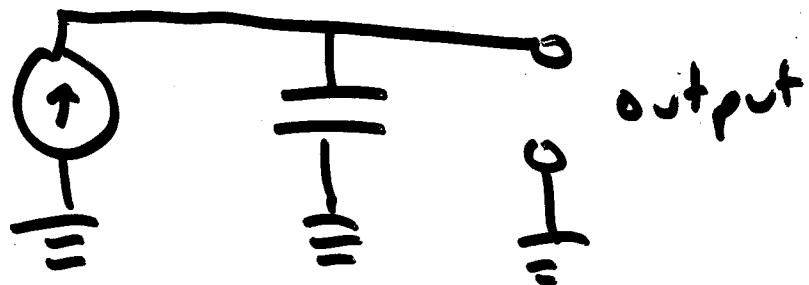


Actually picks off the start of exp. charging curve.

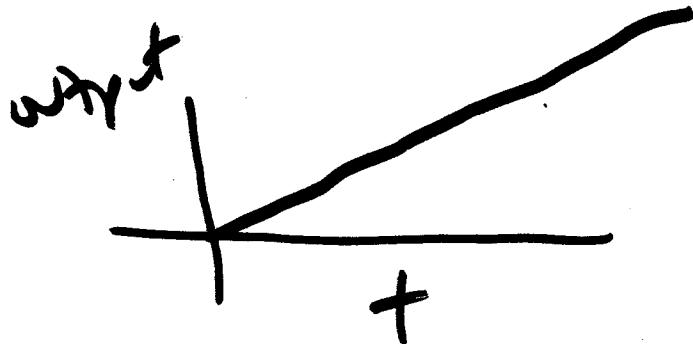
Integrator (cont.)



Ramp Generator: charge a capacitor with constant current.



$$I = C \frac{dV}{dt} \Rightarrow V(t) = \left(\frac{I}{C}\right)t$$



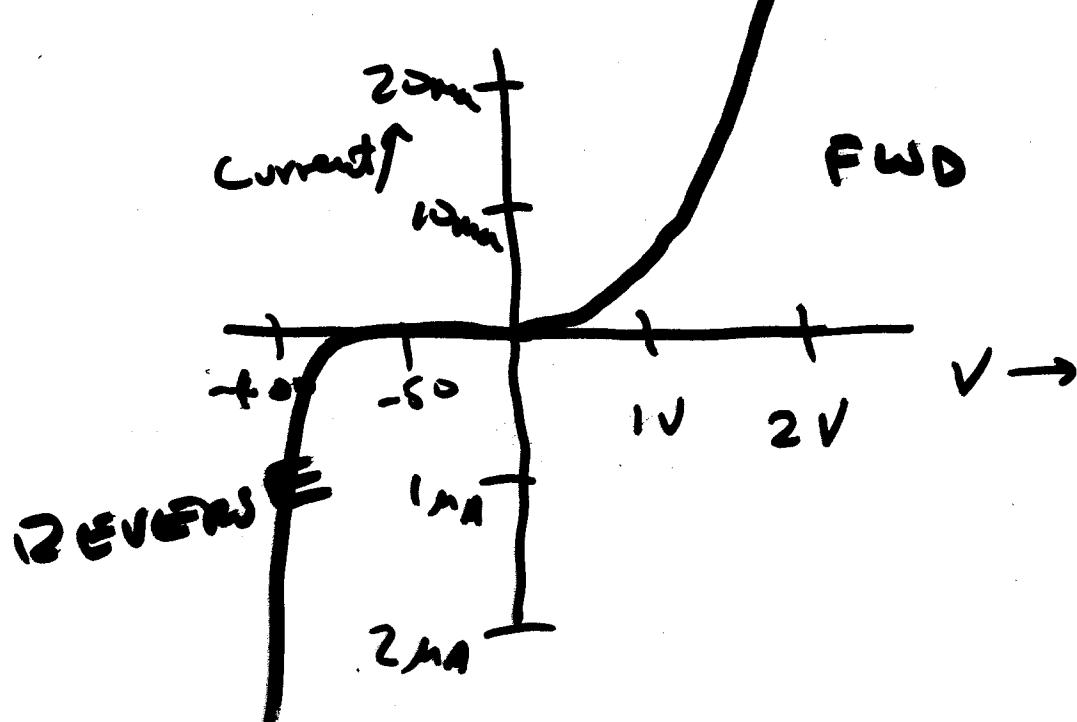
Diodes

R, L are linear circuit elements. (Double signals double the response.)

Also they are 2-terminal, passive (no power).



Diode is non-linear, passive, 2-terminal

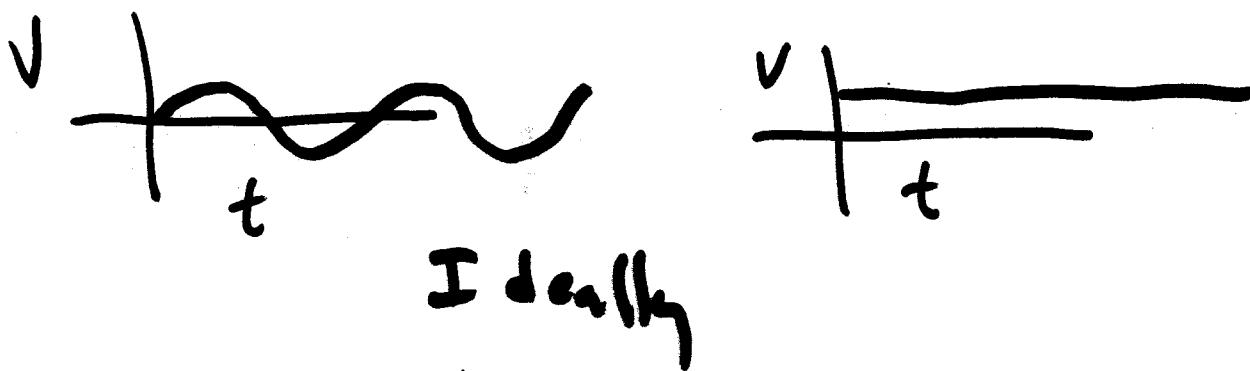


Breakdown

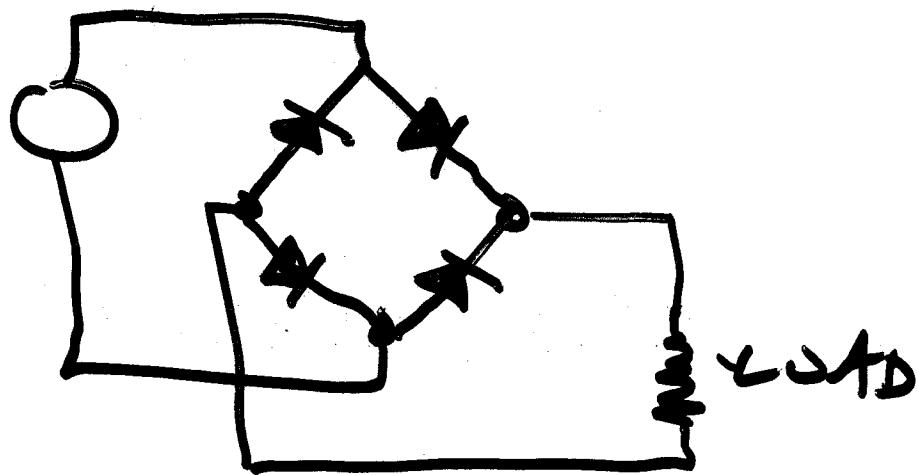
Diodes \rightarrow

- Anode (arrow) points in dir. of forward current flow.
- Approximates 1-way conductor.
- Voltage drop of .5 - .8 V.
- BUT no resistance in terms of Ohm's Law.

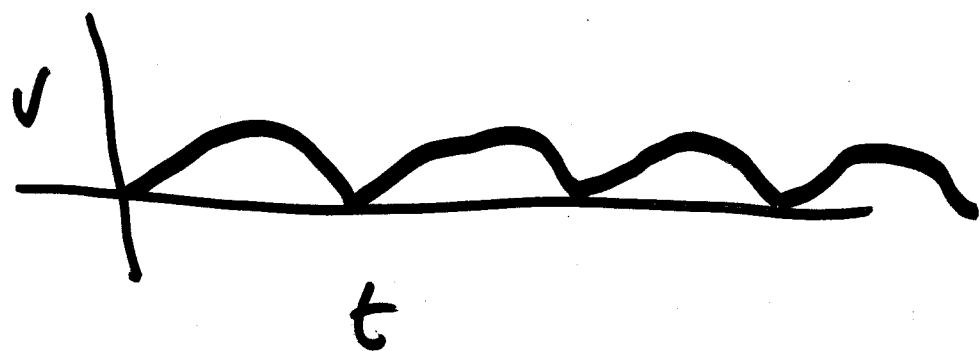
E.g. Rectification
(AC \rightarrow DC)



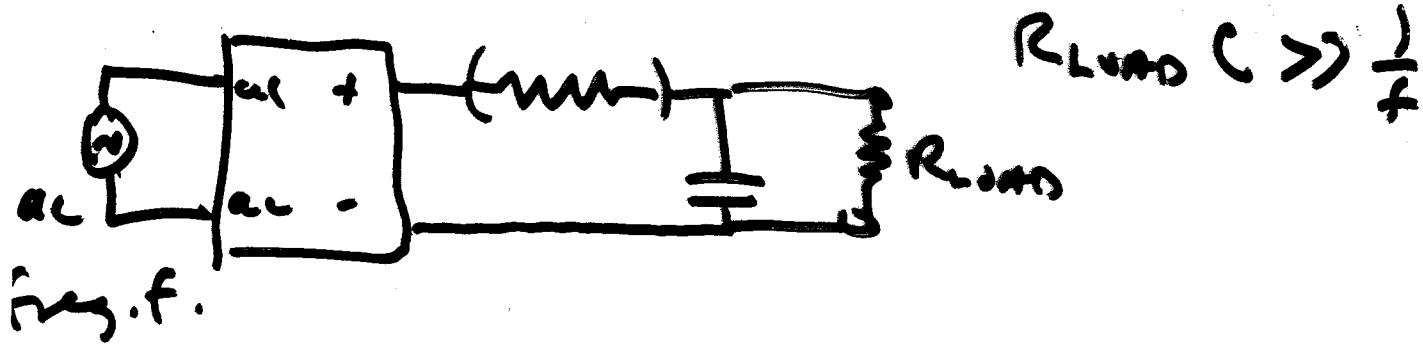
Full Wave Rectifier.



2 Diodes in series = twice the voltage drop. Bad for low-voltage supplies.

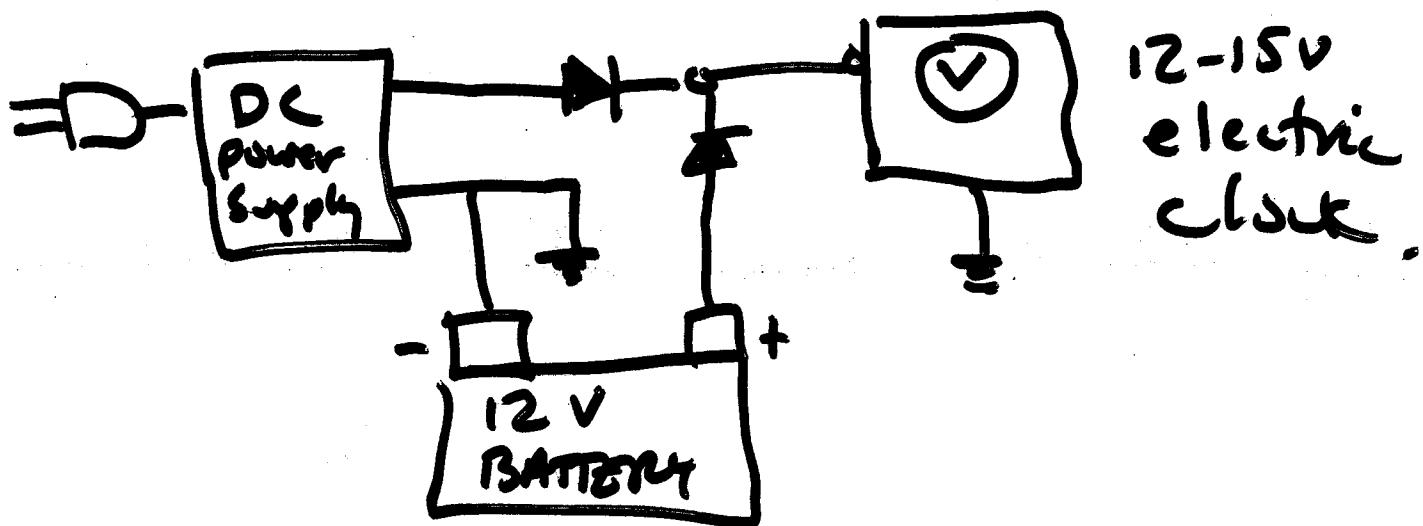


Mega-Cap: FILTERING.
or ... How smooth this out?



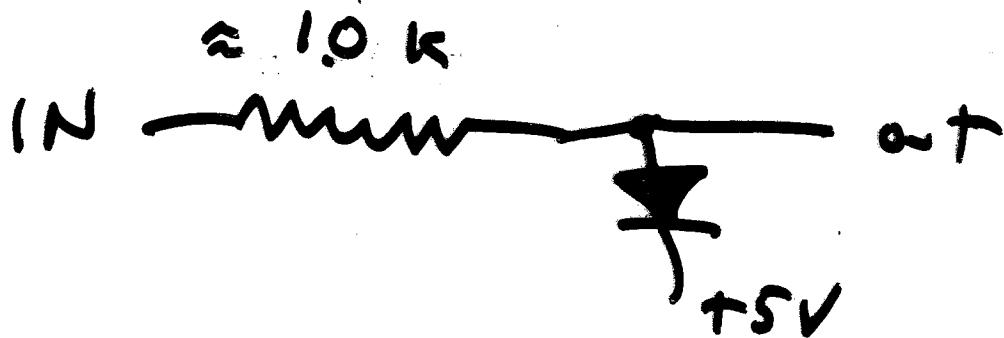
E.g. Diode Gate

Poss Higher of 2 voltages



E.g. Diode Clamp

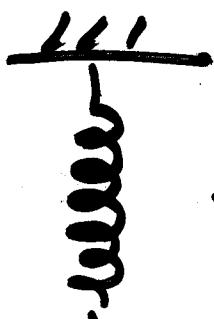
Limit voltage in a signal to ≈ 5.6 V, with no effect on lesser (including negative) voltages. (Protect CMOS from static)



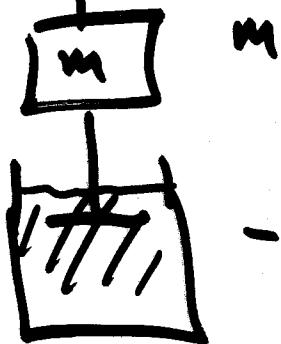
Damped Harmonic Oscillator

- 2nd-order system

- Very basic approach to all systems.



$-kx$ spring force

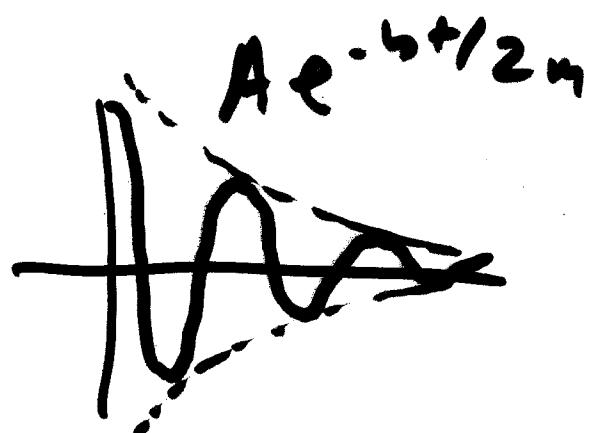


$-b \frac{dx}{dt}$ damping force

$$F = ma$$

$$-kx - b \frac{dx}{dt} = m \frac{d^2x}{dt^2}$$

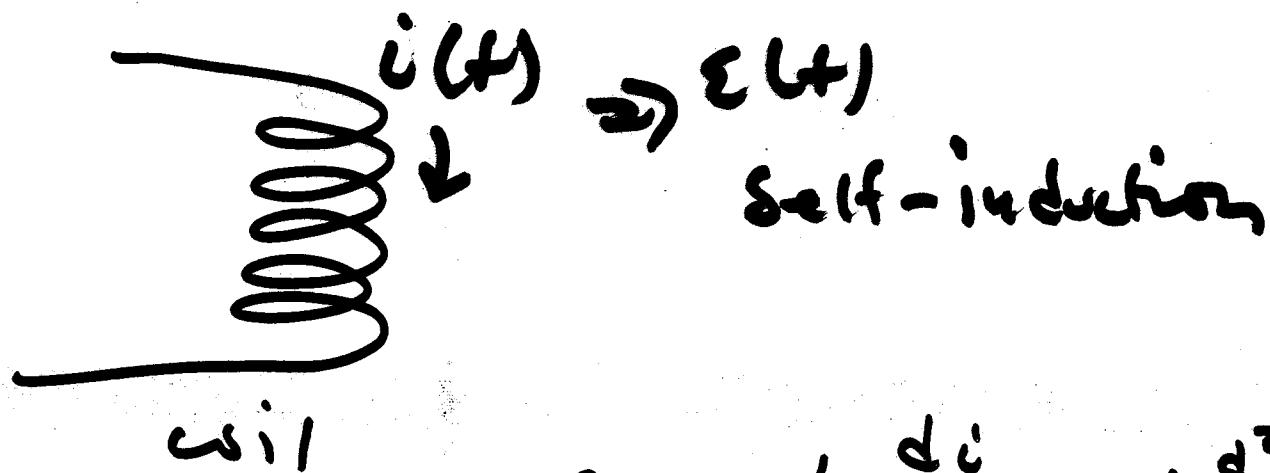
$$m \frac{d^2x}{dt^2} + b \frac{dx}{dt} + kx = 0$$



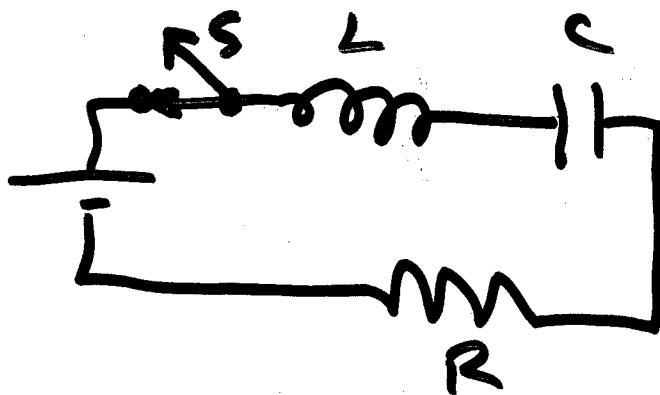
$$x = Ae^{-bt/2m} \cos(\omega' t + \delta)$$

$$\omega' = \sqrt{\frac{k}{m} - \left(\frac{b}{2m}\right)^2}$$

INDUCTANCE



$$\varepsilon = -L \frac{di}{dt} = -L \frac{d^2q}{dt^2}$$



$$L \frac{d^2q}{dt^2} + R \frac{dq}{dt} + \frac{1}{C} q = 0 \quad \}$$

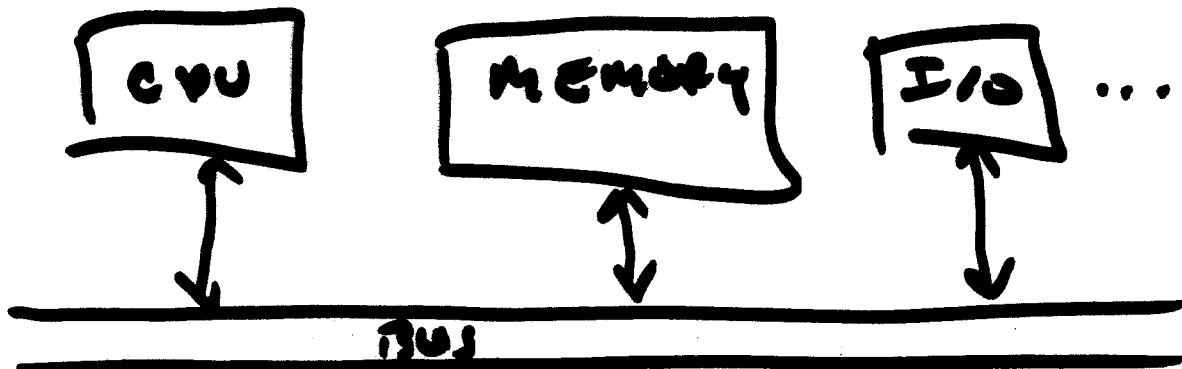
$$m \frac{d^2x}{dt^2} + b \frac{dx}{dt} + kx = 0 \quad \}$$

⇒ Analog computers, controllers, . . .

MicroComputers 101 (J&F ch. 3)

IC hides many realities. It is a multi-tasking OS. But not a bad idea to learn bare-bones how concept. (CSC 254?).

μ controllers come with useful capabilities for interfacing: serial lines, A/D, timers, pulse counters. Replaces multi-card system.



compare to RW's MC684811

p. 51

CPU clock:

XTAL

instruction timing 2 - 40 cycles
(1 - 20.5 μ sec.)

instr: fetch
decode
fetch operands
operate } Bus

Bus is 16 bits wide so 2^{16} ; 64K
locations. (8-bit bytes).

fig 3.5 shows how bus is configured
Shows how address + data live on it.

ports look like memory to CPU
(locations).

to outside, ports are lines to take
signals.

Memory: RAM, ROM, PROM, EEPROM
EEPROM

μ Controller versions exist with all memory types ...